

Optimal Steady-State Control

(with Application to Secondary Frequency Control of Power Systems)

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UNIVERSITY OF
WATERLOO

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Collaborators



Liam S. P. Lawrence
MASc 2019, Waterloo



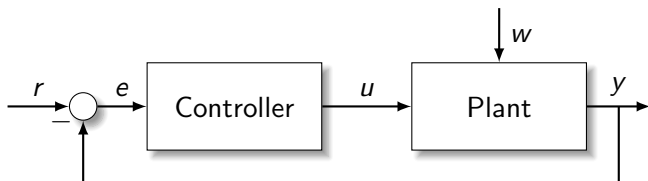
Enrique Mallada
John's Hopkins Univ.

Talk based on:

- 1 *Lawrence, JWSP, Mallada: The optimal steady-state control problem (Arxiv preprint, revision pending ...)*

Control Systems 101

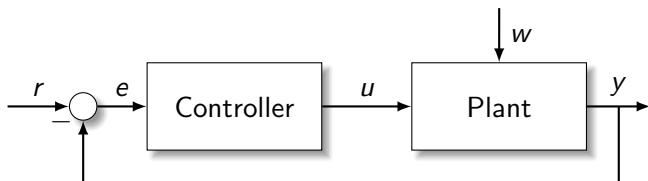
- Prototypical feedback control problem is **tracking** and **disturbance rejection** in the presence of **model uncertainty**



How is the reference r being determined?

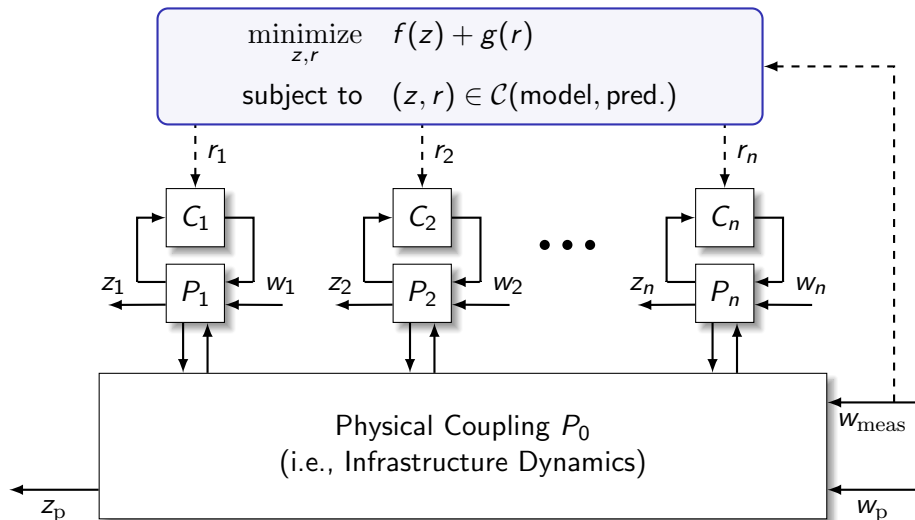
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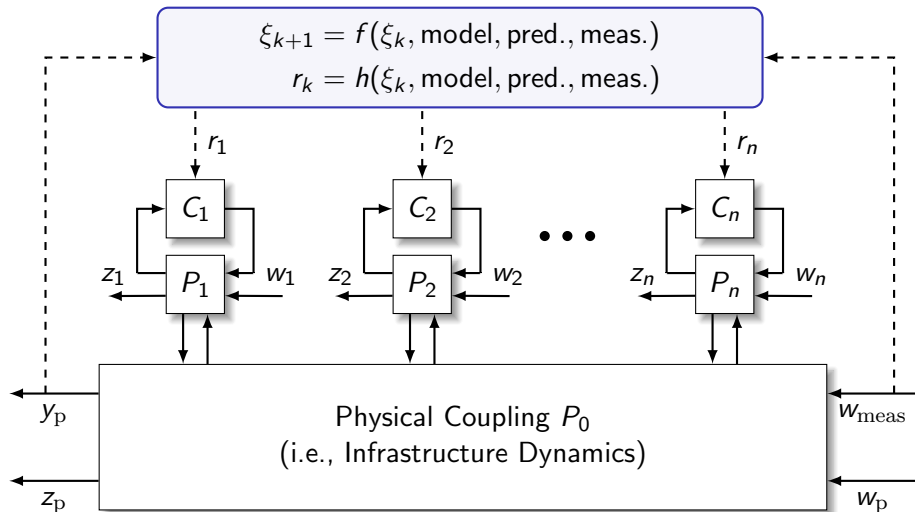


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Feedforward Optimization of Large-Scale Systems



Feedback Optimization of Large-Scale Systems



Optimal Steady-State Control Problem Statement

Given:

- 1 a dynamic system model with
 - a *class* of external disturbances $w(t)$
 - a model uncertainty specification (e.g., parametric)
- 2 a vector of outputs $y \in \mathbb{R}^p$ of system to be optimized
- 3 an optimization problem in y

Design, if possible, a feedback controller such that

- 1 closed-loop is (robustly) well-posed and internally stable
- 2 the regulated output tracks its optimal value

$$\lim_{t \rightarrow \infty} y(t) - y^*(t) = 0, \quad \forall \underline{\text{disturb}}, \forall \underline{\text{uncertainties}}$$

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LTI-Convex OSS Control: Setup Overview

1 Uncertain (possibly unstable) LTI dynamics

$$\begin{aligned}\dot{x} &= A(\delta)x + B(\delta)u + B_w(\delta)w \\ y_m &= C_m(\delta)x + D_m(\delta) + Q_m(\delta)w \\ y &= C(\delta)x + D(\delta)u + Q(\delta)w\end{aligned}$$

- δ = parametric **uncertainty**, w = const. **disturbances**
- y_m = system measurements available for **feedback**
- y = system states/inputs to be **optimized**

2 a steady-state **convex optimization problem**

$$y^*(w, \delta) = \underset{y \in \mathbb{R}^p}{\operatorname{argmin}} \{ f(y, w) : y \in \mathcal{C}(w, \delta) \}$$

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LTI-Convex OSS Control: Setup I

Forced equilibria $(\bar{x}, \bar{u}, \bar{y})$ satisfy

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This defines an **affine set** of *achievable* steady-state outputs

$$\bar{Y}(w, \delta) = (\text{offset vector}) + V(\delta)$$

Note: Due to

- 1 selection of variables $y \in \mathbb{R}^p$ to be optimized, and/or
- 2 structure of state-space matrices (A, B, C, D)

it may be that $\bar{Y}(w, \delta) \subset \mathbb{R}^p$

constraint $\bar{y} \in \bar{Y}(w, \delta)$ cannot be ignored!!

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LTI-Convex OSS Control: Setup II

Desired regulated output $y^*(w, \delta)$ solution to

minimize	$f_0(y, w)$	(convex cost)
	$y \in \mathbb{R}^p$	
subject to	$y \in \bar{Y}(w, \delta)$	(equilibrium)
	$Hy = Lw$	(engineering equality)
	$Jy \leq Mw$	(engineering inequality)

Equilibrium constraints ensure **compatibility** between the plant and the optimization problem

\implies guarantees a steady-state exists s.t. $y = y^*(w, \delta)$.

We want to **track** optimal output $y^*(w, \delta)$

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Optimality Models for OSS Control

An **optimality model** filters the available measurements to robustly produce a proxy error ϵ for the true tracking error $e = y^*(w, \delta) - y$

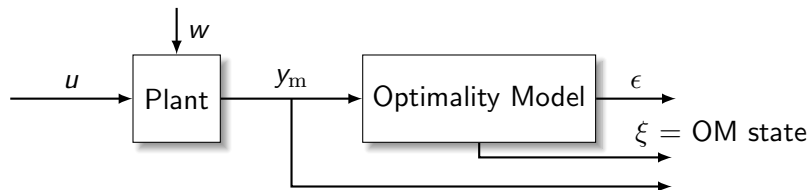


Steady-state requirement: if the plant and optimality model are both in equilibrium and $\epsilon = 0$, then $y = y^*(w, \delta)$.

Driving ϵ to zero (+ internal stability) drives y to y^*

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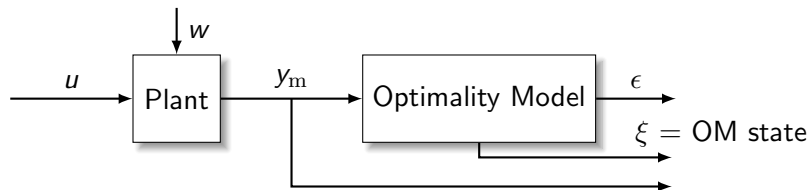


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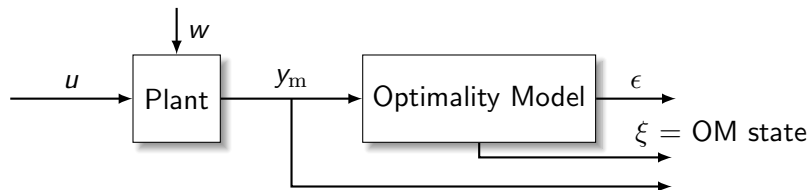


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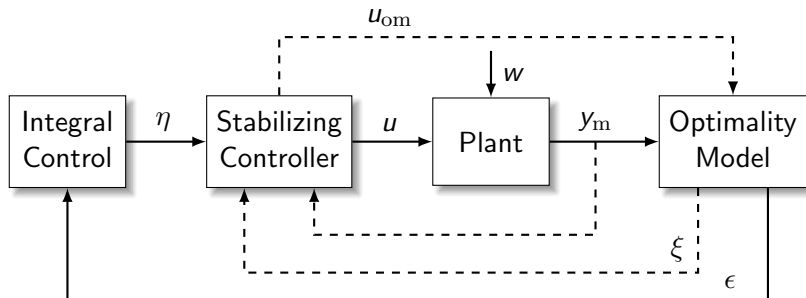


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Big Picture for OSS Control

Optimality model reduces OSS control to output regulation



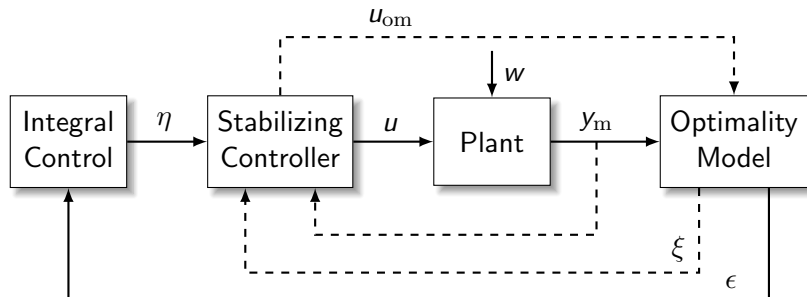
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Integral Control: integrates ϵ

Stabilizing Controller: stabilizes closed-loop system

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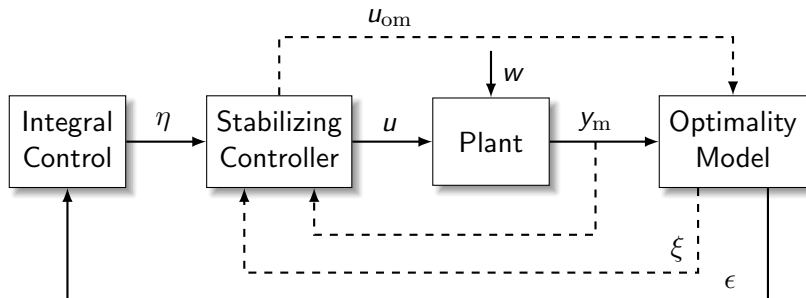
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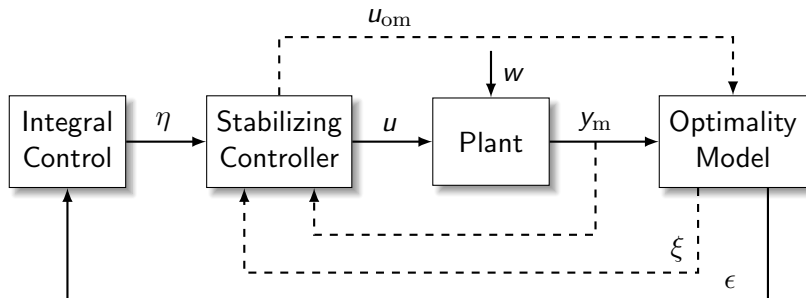
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Can we implement an optimality model that is *robust* against δ ?

$$\begin{aligned} & \underset{y \in \mathbb{R}^p}{\text{minimize}} && f_0(y, w) \\ & \text{subject to} && y \in \bar{Y}(w, \delta) = (\text{offset}) + V(\delta) \\ & && Hy = Lw \\ & && Jy \leq Mw \end{aligned}$$

Optimality condition:

$$\nabla f_0(y^*, w) + J^T \nu^* \perp (V(\delta) \cap \text{null}(H))$$

possibly depends on uncertain parameter δ .

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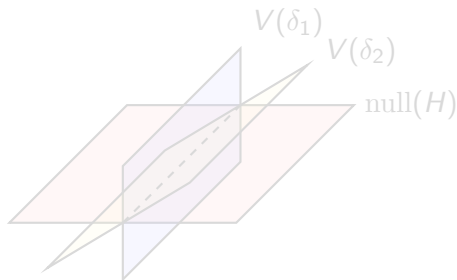
Optimality Model Details II

When can an optimality model encode the gradient KKT condition?

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Robust Feasible Subspace Property

$V(\delta) \cap \text{null}(H)$ is independent of δ



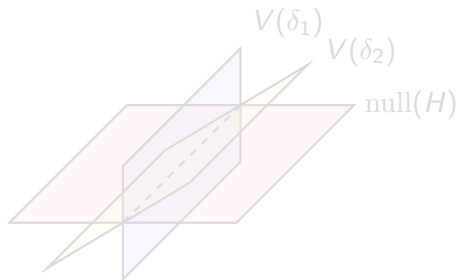
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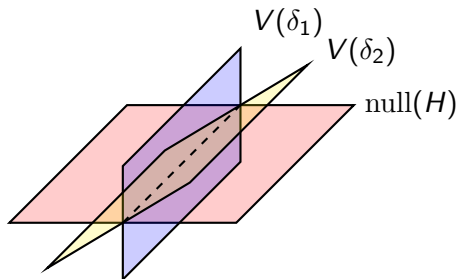
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Optimality Model Details III

If Robust Feasible Subspace property holds, then

$$\begin{aligned} \dot{\nu} &= \varphi(\nu, Jy - Mw) \\ \epsilon &= \begin{bmatrix} Hy - Lw \\ \mathbf{T}_0^T (\nabla f_0(y, w) + J^T \nu) \end{bmatrix} \end{aligned}$$

$$\begin{aligned} &\text{range}(\mathbf{T}_0) \\ &= \mathcal{V}(\delta) \cap \text{null}(H) \end{aligned}$$

(Design freedom!)

is an optimality model for the LTI-Convex OSS Control Problem.

Comments:

- 1 $\mathbf{T}_0^T z$ extracts component of z in subspace $\mathcal{V}(\delta) \cap \text{null}(H)$:

$$\epsilon_2 = 0 \quad \iff \quad \nabla f_0(y, w) + J^T \nu \perp \mathcal{V}(\delta) \cap \text{null}(H)$$

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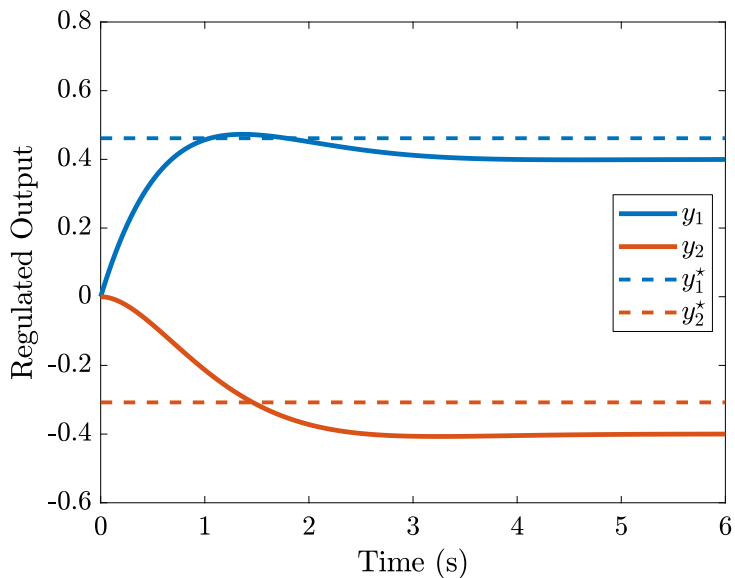
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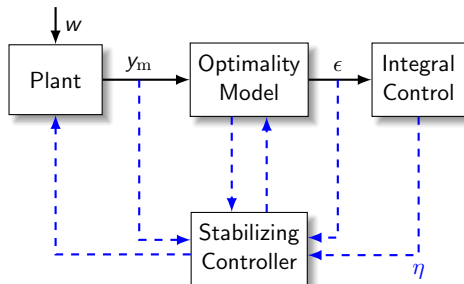
What happens if RFS does not hold?



Stabilizer Details I

Can we actually stabilize this thing?

- 1 **Goal:** stabilize *cascade* of plant \rightarrow optimality model \rightarrow integrators

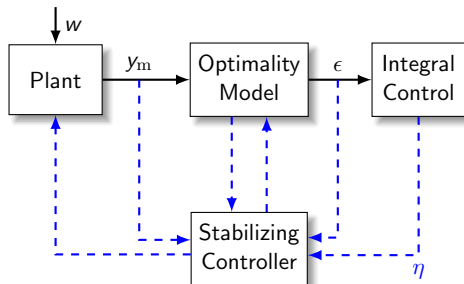


- 2 Can prove closed-loop stable \implies OSS control problem solved
- 3 For QP OSS control, can prove cascade is stabilizable iff
 - plant stabilizable/detectable
 - optimization problem has a unique solution
 - engineering constraints not *redundant* with equilibrium constraints
 - T_0 has full column rank

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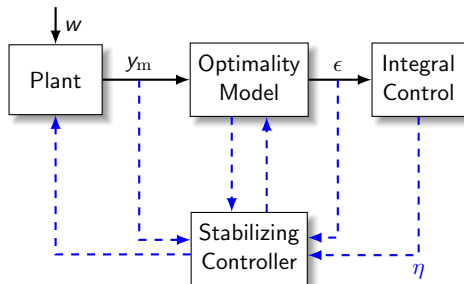


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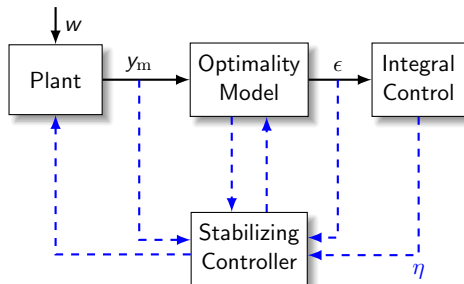


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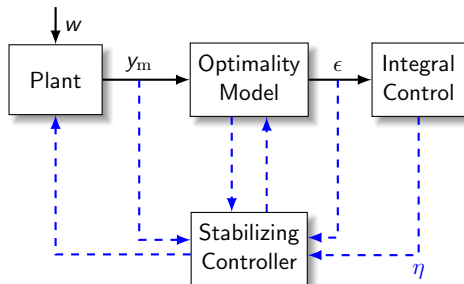


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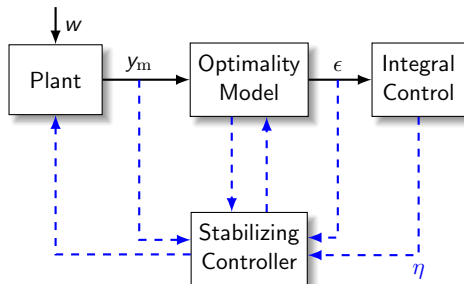


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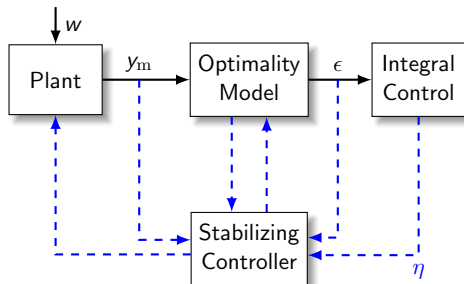


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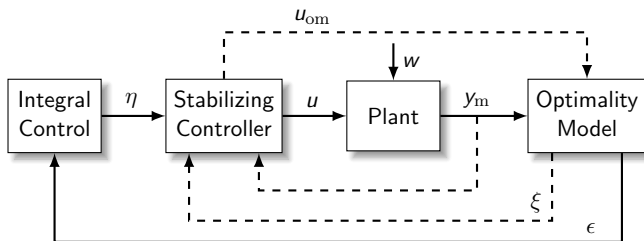
- 1 **Goal:** stabilize *cascade* of plant \rightarrow optimality model \rightarrow integrators



- 2 Can prove closed-loop stable \implies OSS control problem solved
- 3 For QP OSS control, can prove cascade is stabilizable iff
 - plant stabilizable/detectable
 - optimization problem has a unique solution
 - engineering constraints not *redundant* with equilibrium constraints
 - T_0 has full column rank

Stabilizer Details II

Optimality model contains monotone nonlinearity $\nabla f_0(y) \dots$



Stabilizer design options:

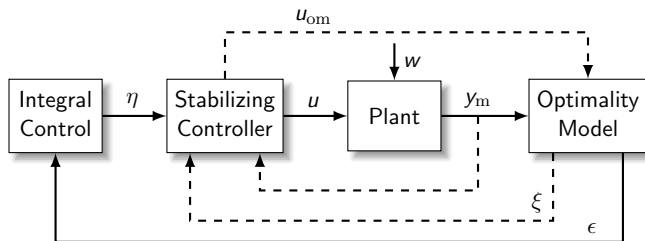
- 1 In *theory*: full-order robustly stabilizing controller design
- 2 In *practice*: low-gain integral control $u = -K\eta$ if open-loop stable, or any heuristic, e.g., linearize and do \mathcal{H}_2 design

Closed-loop stability analysis:

- 1 Robust stability (e.g., IQC-based) or time-scale separation

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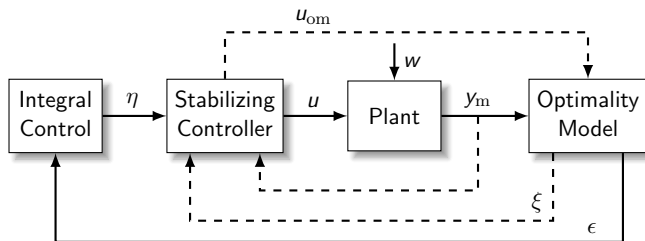
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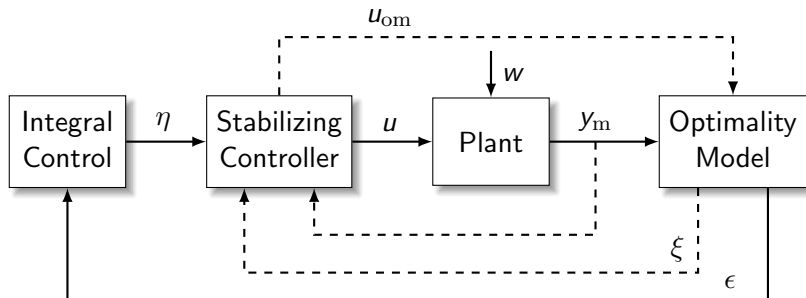
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Big Picture for OSS Control

Optimality model reduces OSS control to output regulation



Optimality Model: creates proxy error signal ϵ

Integral Control: integrates ϵ

Stabilizing Controller: stabilizes closed-loop system

Application: Inexact Reference Tracking

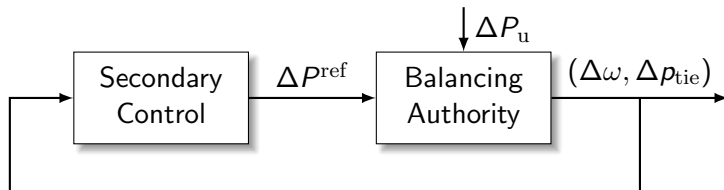
- Want **minimum error** asymptotic tracking of a (possibly infeasible) reference signal subject to actuator limits, e.g.

$$\begin{aligned} & \underset{y_m, u}{\text{minimize}} && \|y_m - r\|_\infty \\ & \text{subject to} && (y_m, u) \in \bar{Y}(w, \delta) \\ & && \underline{u} \leq u \leq \bar{u} \end{aligned}$$

- If reference **feasible**, then exact tracking possible
- Could promote sparsity in steady-state control actions

Application: Frequency Control of Power Systems

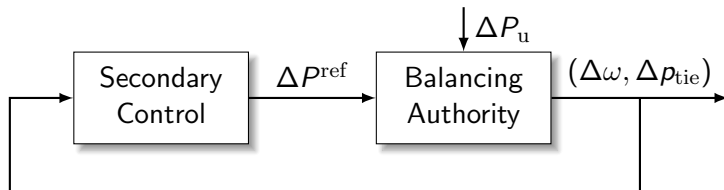
- 1 **Regulate** frequency of an interconnected AC power system in presence of unknown disturbances (locally balance supply and demand)



- 2 Modern challenges / opportunities:
 - variation due to RES \implies need fast control
 - inverter-based resources \implies fast actuation
 - new sensing, comm., comp. \implies new architectures

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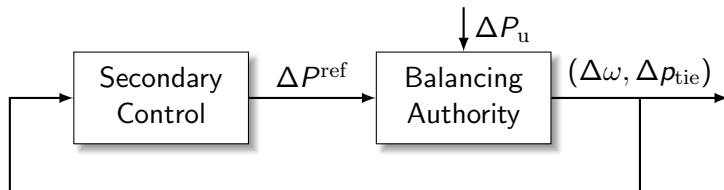
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Key insights into frequency control problem

- 1 For discussion, small-signal network of machines + turbine/gov

$$\begin{aligned}\Delta\dot{\theta}_i &= \Delta\omega_i, \\ M_i\Delta\dot{\omega}_i &= -\sum_{j=1}^n T_{ij}(\Delta\theta_i - \Delta\theta_j) - D_i\Delta\omega_i + \Delta P_{m,i} + \Delta P_{u,i} \\ T_i\Delta\dot{P}_{m,i} &= -\Delta P_{m,i} - R_{d,i}^{-1}\Delta\omega_i + \Delta P_i^{\text{ref}}.\end{aligned}$$

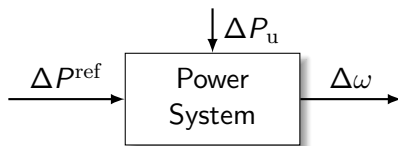
- 2 Model internally stable, DC gain analysis yields

$$\Delta\omega_{\text{ss}} = \frac{1}{\beta} \sum_i \left[\Delta P_i^{\text{ref}} + \Delta P_{u,i} \right]$$

where $\beta = \sum_i (D_i + R_{d,i}^{-1})$ is frequency stiffness.

- 3 Lots of **flexibility** in choice of ΔP^{ref} !

Optimal Allocation of Secondary Resources

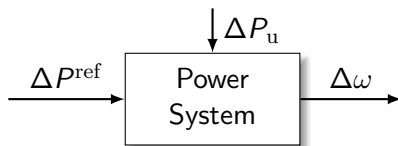


Allocate reserves ΔP_i^{ref} subject to frequency regulation

$$\begin{aligned} & \underset{\Delta P^{\text{ref}} \in \mathbb{R}^n}{\text{minimize}} && \sum_{i=1}^n C_i(\Delta P_i^{\text{ref}}) \\ & \text{subject to} && F \Delta \omega = 0 \end{aligned}$$

- This OSS problem satisfies the robust feasible subspace property \implies can construct (several) different optimality models!

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OSS Framework Recovers Recent Controllers

1 Distributed Averaging PI Control

$$\begin{aligned}\epsilon_i &= \Delta\omega_i - \sum_{j=1}^n a_{ij}(\nabla C_i(P_i^{\text{ref}}) - \nabla C_j(P_j^{\text{ref}})) \\ \dot{\eta}_i &= \epsilon_i \\ P_i^{\text{ref}} &= \text{Stabilizer}_i(\epsilon_i, \eta_i, \omega_i)\end{aligned}$$

- Note: **many** architecture variations possible

2 AGC (stylized version)

$$\dot{\eta} = k \cdot \Delta\omega_{cc}, \quad P_i^{\text{ref}} = (\nabla C_i)^{-1}(\eta)$$

3 Gather-and-broadacst (Dörfler & Grammatico)

$$\dot{\eta} = \frac{1}{n} \sum_{i=1}^n \Delta\omega_i, \quad P_i^{\text{ref}} = (\nabla C_i)^{-1}(\eta)$$

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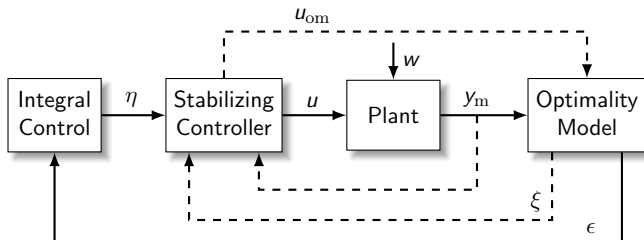
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Conclusions

Optimal Steady-State (OSS) Control framework

- 1 Robust feedback optimization of dynamic systems
- 2 Optimality model reduces OSS problem to output reg. problem



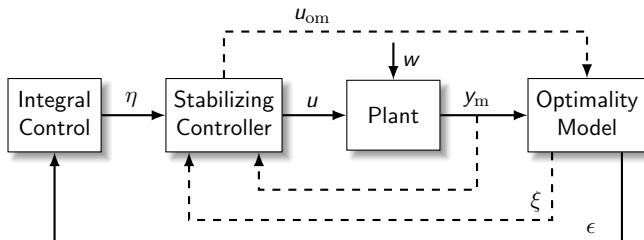
Many pieces of theory wide open ...

- 1 Decentralized, hierarchical, competitive, ...
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The Optimal Steady-State Control Problem

Liam S. P. Lawrence *Student Member, IEEE*, John W. Simpson-Porco, *Member, IEEE*, and Enrique Mallada *Member, IEEE*

<https://arxiv.org/abs/1810.12892>



Liam S. P. Lawrence
University of Waterloo



Enrique Mallada
John's Hopkins Univ.

Questions



<https://ece.uwaterloo.ca/~jwsimpso/>
jwsimpson@uwaterloo.ca

appendix

Feedforward vs. Feedback Optimization

Property	Feedforward	Feedback
Setpoint Quality	\approx Optimal	\approx Optimal
High-Fidelity Model	Crucial	Not crucial
Robustness	No	Yes
Feedback Design/Analysis	Unchanged	More difficult
Computational Effort	Moderate	???

MPC: high computational effort, difficult analysis \Rightarrow Alternatives?

Compared to MPC, if we **give a bit** on *trajectory* optimality, can we can **gain a lot** on ease of design, analysis, and implementation?

Here is a first cut of such an approach.

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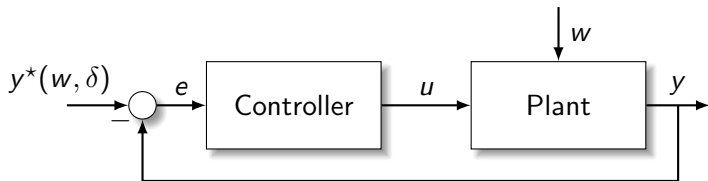
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Is OSS Control just a standard tracking problem?

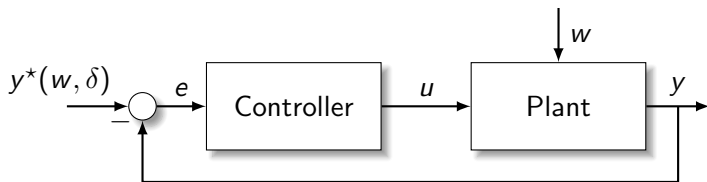


We want y to track $y^*(w, \delta)$, but two problems:

- 1 unmeasured components of w change y^*
- 2 y^* depends on uncertainty δ (relevant if $\bar{Y} \subset \mathbb{R}^P$)

Standard tracking approach **infeasible** for quickly varying $w(t)$, or large uncertainties δ , or particular choices of regulated outputs

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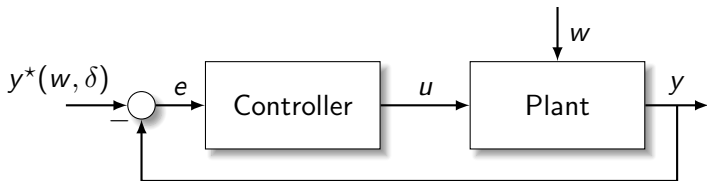


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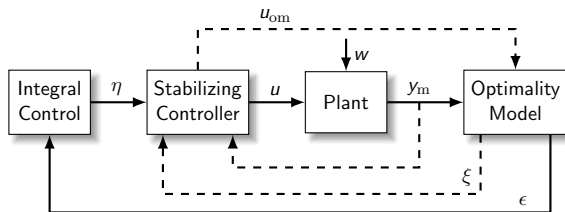
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Towards an internal model principle ...

$$\epsilon = \begin{bmatrix} Hy - Lw \\ T_0^T \nabla f_0(y, w) \end{bmatrix}$$

$$\text{range}(T_0) = \mathcal{V}(\delta) \cap \text{null}(H)$$



Interpretation: Exact robust asymptotic optimization achieved if loop *incorporates a model of the optimal set of the optimization problem*

Slide on EOA Approach . . .

Example 1: Necessity of Equilibrium Constraints

Consider the OSS control problem:

① Dynamics:

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} -1 & 0 \\ 1 & -1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 1 \\ -1 \end{bmatrix} u + \begin{bmatrix} 1 \\ 1 \end{bmatrix} w$$
$$y = \begin{bmatrix} x_1 \\ u \end{bmatrix}$$

② Optimization problem:

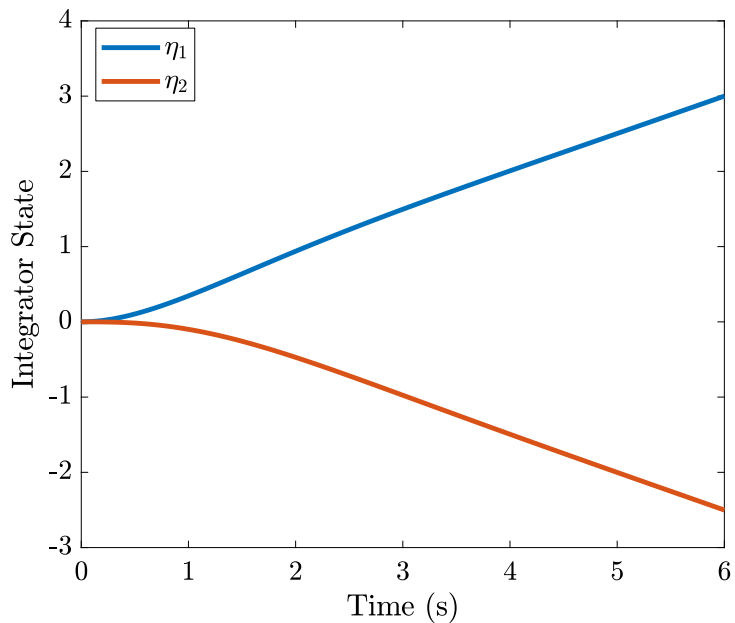
$$\underset{y \in \mathbb{R}^2}{\text{minimize}} \quad g(y) := \frac{1}{2}y_1^2 + \frac{1}{2}y_2^2$$

What happens if we omit the equilibrium constraints?

$$\dot{\eta} = \nabla f_0(y)$$

$$u = -K\eta$$

Example 1: Necessity of Equilibrium Constraints (cont.)



Example 2: Necessity of Robust Feasible Subspace

Consider the OSS control problem:

① Dynamics:

$$\begin{aligned} \begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} &= \begin{bmatrix} -1 - \delta & 0 \\ 1 + \delta & -1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 1 \\ -1 \end{bmatrix} u + \begin{bmatrix} 1 \\ 1 \end{bmatrix} w \\ y &= \begin{bmatrix} x_1 \\ u \end{bmatrix} \end{aligned}$$

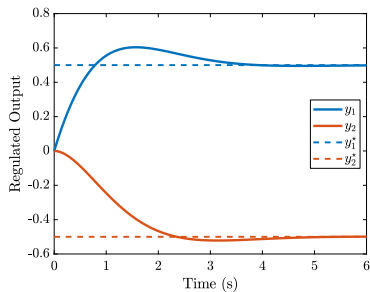
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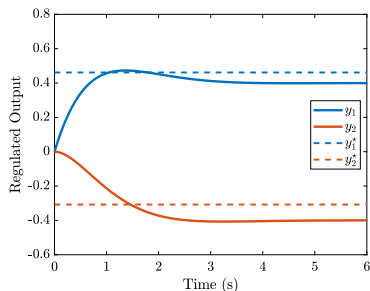
We can show $V(\delta) = \text{span} \left\{ \begin{bmatrix} 1 \\ \delta \end{bmatrix} \right\} \Rightarrow V(\delta)$ **dependent** on δ .

Example 2: Necessity of Robust Feasible Subspace (cont.)

- We apply our scheme anyway supposing $\delta = 0$
- Optimality model + integral control yields...



If $\delta = 0$ in the true plant
 \Rightarrow achieve optimal cost of 0.1538.



If $\delta = 0.5$ in the true plant
 \Rightarrow achieve sub-optimal cost of 0.1599.

Robust Output Subspace Optimality Model

If furthermore $V(\delta)$ itself is independent of δ , then

$$\dot{\mu} = Hy - Lw$$

$$\dot{\nu} = \mathbf{max}(\nu + Jy - Mw, \mathbb{0}) - \nu$$

$$\epsilon = \mathbf{R}_0^T (\nabla f_0(y, w) + H^T \mu + J^T \nu)$$

$$\text{range } R_0 = V(\delta)$$

(Design freedom!)

is also an optimality model for the LTI-Convex OSS Control Problem.

- 1 Can take $R_0 = I$ if $V(\delta) = \mathbb{R}^p$, which holds if

$$\begin{bmatrix} A & B \\ C & D \end{bmatrix} \text{ has full row rank} \iff \text{No transmission zeros at } s = 0$$

- 2 Again, different equivalent formulations of optimization problem give different optimality models

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OSS Control in the Literature

The OSS controller architecture found throughout the literature on **real-time optimization**.

Problem [Nelson and Mallada '18]

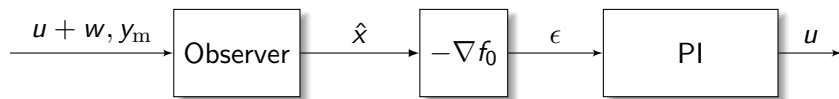
Design a feedback controller to drive the system

$$\begin{aligned}\dot{x}(t) &= Ax(t) + B(u(t) + w) \\ y_m(t) &= Cx(t) + D(u(t) + w)\end{aligned}$$

to the solution of the optimization problem

$$\underset{x \in \mathbb{R}^n}{\text{minimize}} f(x).$$

OSS Control in the Literature (cont.)



Controller Design

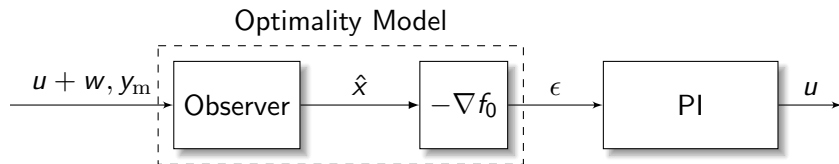
The **optimality model** is an observer with gradient output

$$\begin{aligned}\dot{\hat{x}} &= (A - LC)\hat{x} + (B - LD)(u + w) + Ly_m \\ \epsilon &= -\nabla f_0(\hat{x}).\end{aligned}$$

A PI controller serves as **internal model and stabilizer**

$$\dot{e}_I = \epsilon, \quad u = K_I e_I + K_P \epsilon.$$

OSS Control in the Literature (cont.)



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